The Q Factor of a Second Order System

Consider a second order underdamped system with transfer function

$$G(s) = \frac{\omega_n^2}{s^2 + 2\zeta\omega_n s + \omega_n^2} \,. \tag{1}$$

Notice the (1) was chosen to have unity static gain. This is merely a convenience.

Then
$$P(\omega) = \frac{\omega_n^4}{(\omega_n^2 - \omega^2)^2 + 4\zeta^2 \omega_n^2 \omega^2} = \frac{1}{(1 - (\omega / \omega_n^2)^2 + 4\zeta^2 (\omega / \omega_n)^2)} = \frac{1}{(1 - r)^2 + 4\zeta^2 r}$$
(2)

where we have defined $r \stackrel{\Delta}{=} (\omega / \omega_n)^2$. $P(\omega)$ called the power spectrum associated with (1). It should be clear that $P(\omega)$ is the FRF magnitude-squared.

In order for (2) to have a resonance, there must be a value of *r* that makes the denominator less than 1. This, in turn will place a requirement on the range of ζ values. Setting $(1-r)^2 + 4\zeta^2 r < 1$ gives $r < 2(1-2\zeta^2)$. Since we cannot have r < 0, it follows that that we must have $\zeta < 1/\sqrt{2}$. This is worthy of

Result 1. A second order underdamped system will have no resonance for $\zeta \ge 1/\sqrt{2}$.

In view of this result, in the all of the following development, we will assume that $\zeta < 1/\sqrt{2}$.

We will now find the frequency, ω_{res} at which (2) achieves a maximum. Setting the derivative of (2) equal to zero and solving for r gives $r = 1 - 2\zeta^2$. Hence, (2) is maximum at the frequency:

$$\omega_{res} = \omega_n \sqrt{1 - 2\zeta^2} . \tag{3a}$$

The value of (2) at the frequency (3a) is:

$$P(\omega_{res}) = \frac{1}{4\zeta^2 (1 - \zeta^2)}.$$
 (3b)

The frequency (3a) where the power of the FRF is a maximum is called the *resonance* (or resonant) *frequency*.

The primary focus of these notes is on the Q-factor associated with (1).

Definition 1 The Q factor associated with an underdamped system is defined as $Q \stackrel{\Delta}{=} \omega_{res} / \Delta \omega$, where $\Delta \omega = \omega_2 - \omega_1$. The frequencies $\omega_{1,2}$ satisfy $P(\omega_{1,2}) = 0.5 P(\omega_{res})$. They are called the *half-power* frequencies.

We will now arrive at the expressions for $\omega_{1,2}$. To this end, consider the frequency ω_0 satisfying $P(\omega_0) = 0.5 P(\omega_{res})$. Then from (2) and (3b) we have:

$$\frac{1}{(1-r)^2 + 4\zeta^2 r} = \frac{0.5}{4\zeta^2 (1-\zeta^2)}$$
(4)

where we have set $r \stackrel{\scriptscriptstyle \Delta}{=} (\omega_0 / \omega_n)^2$. From (4) we have:

$$(1-r)^2 + 4\zeta^2 r = 8\zeta^2 (1-\zeta^2).$$
⁽⁵⁾

This, in turn, gives:

$$r^{2} - 2(1 - 2\zeta^{2})r + [1 - 8\zeta^{2}(1 - \zeta^{2})] = 0.$$
(6)

The solution to (6) is:

$$r_{1,2} = (1 - 2\zeta^2) \pm 2\zeta \sqrt{1 - \zeta^2} .$$
(7)

From (7) we arrive at:

$$\omega_0^2 = \omega_n^2 \left[(1 - 2\zeta^2) \pm 2\zeta \sqrt{1 - \zeta^2} \right].$$
(8)

Using (3), (8) becomes:

$$\omega_0^2 = \frac{\omega_{res}^2}{1 - 2\zeta^2} \bigg[(1 - 2\zeta^2) \pm \sqrt{1 - 8\zeta^2 (1 - \zeta^2)} \bigg].$$
(9)

Hence, we arrive at:

$$\omega_{1,2} = \omega_{res} \sqrt{1 \pm \frac{2\zeta}{1 - 2\zeta^2} \sqrt{1 - \zeta^2}} .$$
 (10)

Before we go further, it should be pointed out that in (10) we must have $\omega_1 > 0$. It is easy to show that this requires that

$$\zeta \le 0.38265$$
. (11)

Hence, we see that it is (11) and not $\zeta < 1/\sqrt{2}$ that bounds ζ .

$$\sqrt{r_{1,2}} = \sqrt{1 \pm \frac{2\zeta}{1 - 2\zeta^2} \sqrt{1 - \zeta^2}} \,. \tag{12}$$

We then have

From (7) we have:

Result 2. The Q-factor associated with (1) is: $Q = 1/(\sqrt{r_2} - \sqrt{r_1})$ for $\zeta \le 0.38265$. For $\zeta > 0.38265$, Q is not defined.

Remark 1. Many textbooks on the subject give the approximate expression $\hat{Q} = 1/2\zeta$. The accuracy of this approximation is shown at right.

We see that, indeed, $\hat{Q} = 1/2\zeta$ is an excellent approximation of $Q = 1/(c_2 - c_1)$ for $\zeta \le 0.2$. For $\zeta = 0.38265$, \hat{Q} is ~2.6dB higher than the true value. Many textbooks fail to mention is that it is valid for only $\zeta \le 0.38265$. Moreover, they do not give the expression for the true Q. Fortunately, Q is rarely addressed for damping ratios $\zeta > 0.25$.

[Note: $Q_{dB} = 20 \log Q$.]

 $-20m \, dB/dec$.

The FRF plots below offer some visual appreciation.





Figure 2. Power Spectrum plots for $\zeta = 0.01$ (LEFT), $\zeta = 0.1$ (CENTER) and $\zeta = 0.38268$ (RIGHT).

A Brief Look Into the More General Case with Poles at the Origin

We will now consider the more general case where

One should suspect that this roll-off will values such that the peak at resonance is

the power spectrum associated with (13)

 $P(r) = \frac{1}{r^{m}[(1-r)^{2} + 4\zeta^{2}r]}$

$$G(s) = \frac{\omega_n^{2+m}}{s^m (s^2 + 2\zeta \omega_n s + \omega_n^2)}.$$
(13)
Note that the power 2+min the numerator of (13) is, again,
merely for convenience. Because (13) has m poles at the origin,
its low frequency power behavior will entail a slope of
-20m dB/dec.
One should suspect that this roll-off will reduce the range of ζ -
values such that the peak at resonance is at least 3dB above the $\frac{m}{2}$
value at ω_1 . As was done in relation to (2), let $r \stackrel{\Delta}{=} (\omega / \omega_n)^2$. Then
the power spectrum associated with (13) is:

$$P(r) = \frac{1}{r^m [(1-r)^2 + 4\zeta^2 r]}$$
(14).
(13)

Figure 3. The power spectrum (14) for $\zeta = 0.1$.

Figure 3 shows the power spectrum (14) for $\zeta = 0.1$. The -3dB relative BW is $\Delta r = r_2 - r_1 = 1.07 - 0.85 = 0.22$, and the relative resonance frequency is $r_{res} = 0.98$. Hence, $Q = r_{res} / \Delta r = 4.45$, or $Q_{dB} = 12.97$ dB. This value is indicated by the double arrow in Figure 3. If we approximate Q by $\hat{Q} = 1/2\zeta = 5$, or $\hat{Q}_{dB} = 13.98$ dB, we see that it is about 2dB higher than the true value. It is important to note that the double arrow was obtained by first constructing a straight-line approximation of the power spectrum.

Note from Figure 3 that P(r) has both a relative minimum and a relative maximum. Setting P'(r) = 0 gives:

$$c_{1,2}^{(m)} = \left[(m+1)(1-2\zeta^2) \pm \sqrt{1-4(m+1)^2 \zeta^2 (1-\zeta^2)} \right] / (m+2).$$
(15)

The square root of the smaller element of (15) gives the relative *minimum* scaled frequency prior to the resonance, while the larger element gives the *resonance* scaled frequency.

Proceeding to use (15) in order to arrive at an explicit expression for $P(r_{res})$ would allow us to identify the largest value for ζ such that the Q-factor is defined. Instead, we simply searched for it. The result of this search is shown in Figure 4.

The relative frequencies computed from (15) are $\sqrt{c_1^{(1)}} = 0.594$ and $\sqrt{c_2^{(1)}} = \sqrt{r_{res}^{(1)}} = 0.945$. They match well with 0.6034 and 0.9564, respectively. The damping ratio $\zeta = 0.1475$ results in a 3dB dip between the two frequencies. Hence, $\zeta = 0.1475$ is the largest damping ratio for which a *Q*-factor is defined for m = 1. The true and estimated Q-factors are



Q = 0.9564 / (1.086 - .6034) = 1.98 = 5.94 dB and $\hat{Q} = 1 / (2 \times 0.145) = 3.39 = 10.6 dB$. We see that the estimate is 6.5dB greater than the truth.

While it would have been instructive to arrive at the expression for Q by finding expressions for the -3dB frequencies $\omega_{l}^{(1)}$ and $\omega_{2}^{(1)}$ as was done for m = 0, we did, at least find the expression for the resonance frequency $\omega_{res}^{(m)} = \omega_n \sqrt{c_2^{(m)}}$.

Result 2. For m = 1, the *Q*-factor is not defined for $\zeta \ge 0.1475$. The resonance frequency for any *m* is $\omega_{res}^{(m)} = \omega_n \sqrt{c_2^{(m)}}$.

Conclusions: For second order underdamped systems, the *Q*-factor is commonly taken to be $\hat{Q} = 1/2\zeta$. In fact, in some textbooks it is taken as the definition of the *Q*-factor. If one accepts Definition 1, then:

- For m = 0 $\hat{Q} = 1/2\zeta$ is a reasonable estimate of Q for $\zeta \le 0.25$. A Q-factor only exists for $\zeta \le 0.38$.
- For m=1 $\hat{Q}=1/2\zeta$ is a reasonable estimate of Q for $\zeta \le 0.1$. A Q-factor only exists for $\zeta \le 0.14$.
- For m = 2 one can expect that A *Q*-factor only exists for, at most, $\zeta \leq 0.1$.
- The above bullets are perhaps the likely reason that often engineers do not address a Q-factor for $\zeta \ge 0.1$. I, myself am guilty of having violated this bound B

Matlab Code

```
%PROGRAM NAME: Qfactors.m
%This code investigates two definitions of the Q-factor
%for a 2nd order system:
%Qhat=(2*z)^-1
%Q=wr/dw where dw is the 1/2 power BW range.
8-----
z=0.01:.0001:.38265;
%max z for r1>0 is 0.3826
rl=sqrt(1-2*z.*(1-2*z.^2).^-1.*sqrt(1-z.^2));
sqr=sqrt(1+2*z.*(1-2*z.^2).^-1.*sqrt(1-z.^2));
Q=(sqr-r1).^-1;
Qhat=0.5*z.^-1;
QdB=20*log10(Q);
QhatdB=20*log10(Qhat);
figure(1)
plot(z,QdB,'b','LineWidth',2)
hold on
plot(z,QhatdB,'r--','LineWidth',2)
title('True and Approximate Q-Factors')
legend('Q','Qhat')
ylabel('dB')
xlabel('Damping Ratio (z)')
grid
wn=1000;
z=0.38268;
wr=wn*sqrt(1-2*z^2);
Pwr=1/(4*z^2*(1-z^2));
PwrdB=10*log10(Pwr);
Q1=(2*z)^-1
8-
w1=wr*sqrt(1-2*z*(1-2*z^2)^-1*sqrt(1-z^2));
w2=wr*sqrt(1+2*z*(1-2*z^2)^-1*sqrt(1-z^2));
Pwr1=Pwr/2; Pwr1dB=10*log10(Pwr1);
Pwr2=Pwr1; Pwr2dB=Pwr1dB;
dw = w2 - w1:
Q2=wr/dw
8---
w=0.8*w1:.0001:1.2*w2;
M2=wn^4*((wn^2-w.^2).^2+4*z^2*wn^2*w.^2).^-1;
M2dB=10*log10(M2);
figure(2)
semilogx(w,M2dB)
grid
hold on
plot([wr w1 w2], [PwrdB Pwr1dB Pwr2dB], '*r')
title(['Power Sepctrum for z= ',num2str(z),' .'])
xlabel('Frequency (rad/sec)')
ylabel('dB')
%Include m poles @ the origin:
m=1;
z=0.1475;
r=0.01:.001:100;
P=(r.^m.*((1-r).^2+4*z^2*r)).^-1;
PdB=10*log10(P);
sqr=sqrt(r);
figure(11)
semilogx(sqr,PdB)
title('Power Spectrum P(r)')
xlabel('sqrt(r)=w/wn')
ylabel('dB')
legend('z=0.1475')
grid
§_____
% -3dB relative frequencies:
rctr=((m+1)/(m+2))*1-2*z^2;
dr=sqrt (1-(m+1)^{2*4*z^{2*}(1-z^{2})})/(m+2);
r1=rctr-dr;
rres=rctr+dr;
sqr1=sqrt(r1);
sqrres=sqrt(rres);
[sqr1 sqrres]
```